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KRUMHOLZ &	& MENTLIK		SAUNDERS JR, JOSEPH		
600 SOUTH A' WESTFIELD, I	= :=		ART UNIT	PAPER NUMBER	
,			2614		
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Please find below and/or attached an Office communication concerning this application or proceeding.

The time period for reply, if any, is set in the attached communication.

Office Action Summary		Applicati	on No.	Applicant(s)				
		10/533,6	12	ASADA ET AL.				
		Examine	•	Art Unit				
		Joseph Sa		2614				
Period fo	The MAILING DATE of this communication or Reply	n appears on the	e cover sheet with the d	correspondence ad	ddress			
A SHORTENED STATUTORY PERIOD FOR REPLY IS SET TO EXPIRE 3 MONTH(S) OR THIRTY (30) DAYS, WHICHEVER IS LONGER, FROM THE MAILING DATE OF THIS COMMUNICATION. - Extensions of time may be available under the provisions of 37 CFR 1.136(a). In no event, however, may a reply be timely filed after SIX (6) MONTHS from the mailing date of this communication. - If NO period for reply is specified above, the maximum statutory period will apply and will expire SIX (6) MONTHS from the mailing date of this communication. - Failure to reply within the set or extended period for reply will, by statute, cause the application to become ABANDONED (35 U.S.C. § 133). Any reply received by the Office later than three months after the mailing date of this communication, even if timely filed, may reduce any earned patent term adjustment. See 37 CFR 1.704(b).								
Status								
1)⊠	Responsive to communication(s) filed on 3	10 April 2009						
•	This action is FINAL . 2b) This action is non-final.							
3)	Since this application is in condition for allowance except for formal matters, prosecution as to the merits is							
٠,١	closed in accordance with the practice under <i>Ex parte Quayle</i> , 1935 C.D. 11, 453 O.G. 213.							
Dispositi	on of Claims							
4)⊠	Claim(s) 1-20 is/are pending in the applica	ation.						
-	4a) Of the above claim(s) is/are withdrawn from consideration.							
	5) Claim(s) is/are allowed.							
	6) Claim(s) 1-20 is/are rejected.							
· ·	Claim(s) is/are objected to.							
-	Claim(s) are subject to restriction a	nd/or election r	equirement.					
Applicati	on Papers							
91□	The specification is objected to by the Exar	miner						
•	-		nted or b)□ objected t	o by the Examine	r			
10) The drawing(s) filed on 14 March 2008 is/are: a) accepted or b) objected to by the Examiner.								
Applicant may not request that any objection to the drawing(s) be held in abeyance. See 37 CFR 1.85(a). Replacement drawing sheet(s) including the correction is required if the drawing(s) is objected to. See 37 CFR 1.121(d).								
11) The oath or declaration is objected to by the Examiner. Note the attached Office Action or form PTO-152.								
	ınder 35 U.S.C. § 119							
	-	oian priority up	dor 25 11 C.C. \$ 110/o) (d) or (f)				
•	Acknowledgment is made of a claim for for	eign priority un	der 35 U.S.C. § 119(a)-(a) or (1).				
a)	a) ☐ All b) ☐ Some * c) ☐ None of: 1. ☐ Certified copies of the priority documents have been received.							
	2. Certified copies of the priority documents have been received in Application No							
	3. Copies of the certified copies of the priority documents have been received in this National Stage							
application from the International Bureau (PCT Rule 17.2(a)).								
* See the attached detailed Office action for a list of the certified copies not received.								
Attachmen			_					
1) Notice of References Cited (PTO-892) 4) Interview Summary (PTO-413) 2) Notice of Draftsperson's Patent Drawing Review (PTO-948) Paper No(s)/Mail Date								
2) Notice of Draftsperson's Patent Drawing Review (PTO-948) 3) Information Disclosure Statement(s) (PTO/SB/08) Paper No(s)/Mail Date 5) Notice of Informal Patent Application								
Paper No(s)/Mail Date 6) Other:								

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DETAILED ACTION

This office action is in response to the communications filed April 10, 2009.
 Claims 1 – 20 are currently pending and considered below.

Claim Rejections - 35 USC § 112

- 2. The following is a quotation of the first paragraph of 35 U.S.C. 112:
 - The specification shall contain a written description of the invention, and of the manner and process of making and using it, in such full, clear, concise, and exact terms as to enable any person skilled in the art to which it pertains, or with which it is most nearly connected, to make and use the same and shall set forth the best mode contemplated by the inventor of carrying out his invention.
- 3. Claims 1 20 are rejected under 35 U.S.C. 112, first paragraph, because the specification, while being enabling for "such that the frequency response to the audio signal at a second point in the sound field is lower than the frequency response to the audio signal at the first point in the sound field", does not reasonably provide enablement for "such that the frequency response to the audio signal at the first point in the sound field is lower than the frequency response to the audio signal at a second point in the sound field". The specification does not enable any person skilled in the art to which it pertains, or with which it is most nearly connected, to make and use the invention commensurate in scope with these claims. Applicant points to paragraph [0063] in the Remarks dated April 10, 2009 on page 9, however in paragraph [0060] it states, "Since the delay component of each of the digital filters DL0 to DLn is set for focusing the sound output at the point Ptg, the space synthesis impulse response Itg measured at the point Ptg will be a large impulse as shown in FIG. 1. Also, the frequency response (amplitude part) Ftg of the space synthesis impulse response Itg

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will be flat in the entire frequency band as shown in FIG. 4 because the time waveform takes the form of an impulse. Therefore, the sound pressure will be increased at the point Ptg.," emphasis added by Examiner. And in paragraph [0063] it states, "At this, since the space synthesis impulse response Itg at the sound pressure increasing point Ptg is a large impulse while the space synthesis impulse response Inc at the point Pnc is a signal having dispersed impulses, the frequency response Fnc at the point Pnc will be lower in level than the frequency response Ftg at the point Ptg. Therefore, the sound pressure will be decreased at the point Pnc," emphasis added by Examiner. Therefore since the "first point" as claimed is where the audio signals arrive in the sound field via a plurality of digital filters and each of the plurality of speakers so as to coincide with each other. The "first point" as claimed corresponds to the "point" Ptg" as explained in the specification. That leaves the claimed "second point" to correspond to "point Pnc". For examination purposes the Examiner will interpret the claim limitation as "such that the frequency response to the audio signal at a second point in the sound field is lower than the frequency response to the audio signal at the first point in the sound field". Appropriate correction and clarification is required.

Claim Rejections - 35 USC § 103

- 4. The following is a quotation of 35 U.S.C. 103(a) which forms the basis for all obviousness rejections set forth in this Office action:
 - (a) A patent may not be obtained though the invention is not identically disclosed or described as set forth in section 102 of this title, if the differences between the subject matter sought to be patented and the prior art are such that the subject matter as a whole would have been obvious at the time the invention was made to a person having ordinary skill in the art to which said subject matter pertains. Patentability shall not be negatived by the manner in which the invention was made.

5. Claims 1 – 4 and 10 – 13 are rejected under 35 U.S.C. 103(a) as being unpatentable over Bienek et al. (WO 02/078388 A2), hereinafter <u>Bienek</u>, in view of North (US 6,801,631), hereinafter <u>North</u>.

Claim 1: Bienek discloses an audio signal processing method (method and apparatus to create a sound field) comprising the steps of: supplying an audio signal (input signal 101) to each of a plurality of digital filters (means 1506 includes signal delay means 1508, amplitude control means 1510, and adjustable digital filter 1512); respectively supplying outputs from the plurality of digital filters to a plurality of speakers arranged in a speaker array to form a sound field (Description of Figure 6, Pages 18 – 19); setting a predetermined delay time in each of the plurality of digital filters so that transmission delay times with which the audio signal arrives at a first point in the sound field via each of the plurality of digital filters and each of the plurality of speakers will coincide with each other (Third Sound Field, Pages 21 – 22 and Figure 7C and Figure 8); and adjusting at least one amplitude characteristic of the plurality of digital filters ("The amplitude control means (ACM) is conveniently implemented as digital amplitude control means for the purposes of gross beam shape modification," Page 12 Line 25 – Page 13 Line 6).

Bienek does not explicitly state the amplitude control means is adjusted such that the frequency response to the audio signal at the first point [will be interpreted for examination purposes as "a second point" as explained in the 35 U.S.C. 112 rejection above] in the sound field is lower than the frequency response to the audio signal at a

second point [will be interpreted as "the first point" as explained in the 35 U.S.C. 112 rejection above] in the sound field. Bienek does explain, "The amplitude control means is preferably arranged to apply differing amplitude control to each signal output from the Distributor so as to counteract for the fact that the DPAA is of finite size by using a window function," Page 12 Line 25 – Page 13 Line 6 and Figures 11A – 11D, "The window function reduces the effects of "side lobes" at the expense of power. The type of window function used is chosen dependent on the qualities required of the resultant beam. Thus, if beam directivity is important, a window function as is shown in Figure 11A should be used. If less directivity is important, a window function as is shown in Figure 11D can be used," Page 26 Lines 20 – 25, "Thus, in general, output signals destined for SETs near the centre of the array will not be significantly affected but those near to the perimeter of the array will be attenuated according to how near to the edge of the array they are," Page 12 Line 25 – Page 13 Line 6.

Again, while not explicitly stated by <u>Bienek</u>, <u>North</u> discloses a similar speaker system with multiple transducers positioned in a plane for optimum acoustic radiation pattern and illustrates in Figures 20 – 23 that the frequency response to the audio signal at a second point in the sound field (off-axis) is lower than the frequency response to the audio signal at the first point in the sound field (on-axis). <u>North</u> explains "conventional arrays direct a portion of their sound towards the listener (main lobe) and a portion of their sound towards the sides (side lobes). When placed in a room, the sound waves in the side lobes reflect from surfaces such as walls, the ceiling, and the floor (room reflections). The reflected sound waves interact either constructively or

destructively, depending on the delay time and frequency, with the direct sound waves. The listener is presented with a sound field that is a combination of the direct sound from the main lobe and the reflected and delayed sound from the side lobes. Although the speaker system may have exhibited a flat frequency response in an anechoic chamber, the frequency response in the room at the listener is anything but flat, with pronounced variations in the response. The result is the listener hears a severely distorted sound field that bears little resemblance to the original event," Column 2 Lines 22 – 52. North further illustrates on-axis and off-axis frequency response "waterfall" plots for different speaker arrays (Figures 20 – 23). North explains, "A waterfall plot illustrates the sound dispersion characteristics of a speaker system by cascading the multiple frequency response curves of a speaker system taken at multiple angles. The waterfall plots of FIGS. 20 through 23 represent 10 frequency response measurements taken in 10-degree increments from 0 degrees (on-axis) to 90 degrees off-axis. The measurement curves cascade diagonally on top of one another, starting with the 0degree on-axis curve (the rearmost curve) and ending with the 90-degree off-axis curve (the curve at the front), Column 6 Lines 34 – 44. North further explains, "with less energy directed towards the room walls, floor, and ceiling, the intensity of the room reflections decreases. This decreases the level of sounds that echoes within the room. A decreased level of room reflections can also result in a flatter frequency response in a room as the attenuated reflected waves interfere less with the sound waves in the main lobe," Column 7 Lines 32 – 45.

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Thus in all cases, North illustrates (Figures 20 - 23) that the frequency response to the audio signal at a second point in the sound field (off-axis) is lower than the frequency response to the audio signal at the first point in the sound field (on-axis), especially Figure 22, where North illustrates that a significant reduction to off-axis sound energy due to increased directivity results "in a flatter frequency response in a room as the attenuated reflected waves interfere less with the sound waves in the main lobe," Column 7 Lines 32 - 45.

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Therefore, it would have been obvious to one of ordinary skill in the art at the time of the invention given the teachings of North to utilize the reduction in the effects of "side lobes" by using the amplitude control means by applying a window function to the audio signal as disclosed by Bienek to achieve a frequency response to the audio signal at a second point in the sound field (off-axis) that is lower than the frequency response to the audio signal at the first point in the sound field (on-axis) since it is beneficial to have a significant reduction to off-axis sound energy due to increased directivity given it results "in a flatter frequency response in a room as the attenuated reflected waves interfere less with the sound waves in the main lobe," Column 7 Lines 32 – 45.

Claim 2: <u>Bienek</u> and <u>North</u> disclose the audio signal processing method according to claim 1, wherein a sound wave from the speaker array is caused to reach at least one of the first and second points after it is reflected by a wall surface (<u>Bienek</u> Figure 8).

Claim 3: <u>Bienek</u> and <u>North</u> disclose the audio signal processing method according to claim 1, wherein when forming the first and second points in the sound field, a filter factor of each of the plurality of digital filters is determined by calculation and set for each of the plurality of digital filters (<u>Bienek</u> Third Sound Field, Pages 21 – 22 and Figure 7C).

Claim 4: <u>Bienek</u> and <u>North</u> disclose the audio signal processing method according to claim 1, wherein when forming the first and second points in the sound field, a filter factor of each of the plurality of digital filters is read from a data base and set for each of the plurality of digital filters (stored sets of delays (for the DDGs) and filter coefficients (for the ADFS) can be recalled, Bienek Page 14 Lines 26 and 27).

Claim 10: Bienek discloses an audio signal processor (method and apparatus to create a sound field) comprising a plurality of digital filters (means 1506 includes signal delay means 1508, amplitude control means 1510, and adjustable digital filter 1512) each supplied with an audio signal (input signal 101), wherein each of the plurality of digital filters supplies an output signal to each of a plurality of speakers arranged in a speaker array to form a sound field (Description of Figure 6, Pages 18 – 19); each of the plurality of digital filters has a predetermined delay time so that transmission delay times with which the audio signal arrives at a first point in the sound field via each of the plurality of digital filters and each of the plurality of speakers will coincide with each other (Third Sound Field, Pages 21 – 22 and Figure 7C and Figure 8); and adjusting at least one

amplitude characteristic of the plurality of digital filters ("The amplitude control means (ACM) is conveniently implemented as digital amplitude control means for the purposes of gross beam shape modification," Page 12 Line 25 –Page 13 Line 6).

Bienek does not explicitly state the amplitude control means is adjusted such that the frequency response to the audio signal at the first point [will be interpreted for examination purposes as "a second point" as explained in the 35 U.S.C. 112 rejection above] in the sound field is lower than the frequency response to the audio signal at a second point [will be interpreted as "the first point" as explained in the 35 U.S.C. 112 rejection above] in the sound field. Bienek does explain, "The amplitude control means is preferably arranged to apply differing amplitude control to each signal output from the Distributor so as to counteract for the fact that the DPAA is of finite size by using a window function," Page 12 Line 25 – Page 13 Line 6 and Figures 11A – 11D, "The window function reduces the effects of "side lobes" at the expense of power. The type of window function used is chosen dependent on the qualities required of the resultant beam. Thus, if beam directivity is important, a window function as is shown in Figure 11A should be used. If less directivity is important, a window function as is shown in Figure 11D can be used," Page 26 Lines 20 – 25, "Thus, in general, output signals destined for SETs near the centre of the array will not be significantly affected but those near to the perimeter of the array will be attenuated according to how near to the edge of the array they are," Page 12 Line 25 – Page 13 Line 6.

Again, while not explicitly stated by <u>Bienek</u>, <u>North</u> discloses a similar speaker system with multiple transducers positioned in a plane for optimum acoustic radiation

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pattern and illustrates in Figures 20 – 23 that the frequency response to the audio signal at a second point in the sound field (off-axis) is lower than the frequency response to the audio signal at the first point in the sound field (on-axis). North explains "conventional arrays direct a portion of their sound towards the listener (main lobe) and a portion of their sound towards the sides (side lobes). When placed in a room, the sound waves in the side lobes reflect from surfaces such as walls, the ceiling, and the floor (room reflections). The reflected sound waves interact either constructively or destructively, depending on the delay time and frequency, with the direct sound waves. The listener is presented with a sound field that is a combination of the direct sound from the main lobe and the reflected and delayed sound from the side lobes. Although the speaker system may have exhibited a flat frequency response in an anechoic chamber, the frequency response in the room at the listener is anything but flat, with pronounced variations in the response. The result is the listener hears a severely distorted sound field that bears little resemblance to the original event," Column 2 Lines 22 – 52. North further illustrates on-axis and off-axis frequency response "waterfall" plots for different speaker arrays (Figures 20 – 23). North explains, "A waterfall plot illustrates the sound dispersion characteristics of a speaker system by cascading the multiple frequency response curves of a speaker system taken at multiple angles. The waterfall plots of FIGS. 20 through 23 represent 10 frequency response measurements taken in 10-degree increments from 0 degrees (on-axis) to 90 degrees off-axis. The measurement curves cascade diagonally on top of one another, starting with the 0degree on-axis curve (the rearmost curve) and ending with the 90-degree off-axis curve

(the curve at the front), Column 6 Lines 34 – 44. North further explains, "with less energy directed towards the room walls, floor, and ceiling, the intensity of the room reflections decreases. This decreases the level of sounds that echoes within the room. A decreased level of room reflections can also result in a flatter frequency response in a room as the attenuated reflected waves interfere less with the sound waves in the main lobe," Column 7 Lines 32 – 45.

Thus in all cases, North illustrates (Figures 20 - 23) that the frequency response to the audio signal at a second point in the sound field (off-axis) is lower than the frequency response to the audio signal at the first point in the sound field (on-axis), especially Figure 22, where North illustrates that a significant reduction to off-axis sound energy due to increased directivity results "in a flatter frequency response in a room as the attenuated reflected waves interfere less with the sound waves in the main lobe," Column 7 Lines 32 - 45.

Therefore, it would have been obvious to one of ordinary skill in the art at the time of the invention given the teachings of North to utilize the reduction in the effects of "side lobes" by using the amplitude control means by applying a window function to the audio signal as disclosed by Bienek to achieve a frequency response to the audio signal at a second point in the sound field (off-axis) that is lower than the frequency response to the audio signal at the first point in the sound field (on-axis) since it is beneficial to have a significant reduction to off-axis sound energy due to increased directivity given it results "in a flatter frequency response in a room as the attenuated reflected waves interfere less with the sound waves in the main lobe," Column 7 Lines 32 – 45.

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Claim 11: <u>Bienek</u> and <u>North</u> disclose the audio signal processor according to claim 10, wherein a sound wave from the speaker array is caused to reach at least one of the first and second points after it is reflected by a wall surface (Bienek Figure 8).

Claim 12: <u>Bienek</u> and <u>North</u> disclose the audio signal processor according to claim 10, wherein when forming the first and second points in the sound filter, a filter factor of each of the plurality of digital filters is determined by calculation and set for each of the plurality of digital filters (Bienek Third Sound Field, Pages 21 – 22 and Figure 7C).

Claim 13: Bienek and North disclose the audio signal processor according to claim 10, wherein when forming the first and second points in the sound field, a filter factor of each of the plurality of digital filters is read from a data base and set for each of the plurality of digital filters (stored sets of delays (for the DDGs) and filter coefficients (for the ADFS) can be recalled, Bienek Page 14 Lines 26 and 27).

6. Claims 5 – 9 and 14 – 20 are rejected under 35 U.S.C. 103(a) as being unpatentable over <u>Bienek</u> and <u>North</u> in view of Masako et al. (JP-8-191225-A), hereinafter <u>Masako</u>.

Claim 5: <u>Bienek</u> and <u>North</u> disclose the audio signal processing method according to claim 1, but does not explicitly disclose wherein: the predetermined delay time set for at

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least one of the plurality of digital filters is divided into an integer part and decimal part in units of a sampling period of the audio signal; over-sampling an impulse response including a delay time represented by at least the decimal part of the predetermined delay time for a shorter period than a sampling period to provide a sample train and, wherein the sample train is down-sampled to provide pulse-waveform data of the sampling period; and factor data is set for a part to be delayed by the plurality of digital filters based on the pulse- waveform data. Bienek does disclose "the minimum delay possible for a given signal is preferably as small or smaller than Ts, that signal's sample period" and that "most preferably, the smallest incremental change in delay possible for a given digital signal should be no larger than Ts, that signal's sampling period. Otherwise, interpolation of the signal is necessary," Page 12 Lines 17 – 24. Therefore Bienek does disclose a fractional delay and also discloses that a delay filter and an adaptive digital filter may be used. Bienek does not disclose details of how to perform, for example, the interpolation necessary for the fractional delays disclosed above and therefore one would be inclined to look elsewhere for such a teaching. Masako discloses the technique necessary to include a fractional delay (Figure 6 –7 and Paragraph 29 – 33) and discloses that this technique is an effective approach when slight spacing differences influence the felling of the direction of perceived sound. Therefore, it would have been obvious to one of ordinary skill in the art at the time of the invention to include the process of over-sampling an impulse response and then downsampling to provide a pulse-waveform data of the sampling period as disclosed by

<u>Masako</u> for setting the delay coefficients for the plurality of digital filters as disclosed by <u>Bienek</u> and <u>North</u>, thereby providing better perceived spatial resolution.

Claim 6: Bienek, North, and Masako disclose the audio signal processing method according to claim 5, wherein the audio signal is delayed by a part of the predetermined delay time, which is a multiple of the sampling period, by digital delay circuits which operate for the sampling period, while it is being delayed by the remainder of the predetermined delay time, which includes the decimal part by the digital filters (Bienek discloses that the delay time may be a fractional sampling period and also discloses cascading a delay means with adjustable digital filter means that can also apply delays. Therefore given the disclosure of Bienek and the teachings of Masako of how to calculate a finer representation of an impulse response, Bienek, North, and Masako disclose implementing a delay using a simple delay element and an adjustable digital filter for the remainder or fractional part of the delay in a two stage process as disclosed by Bienek).

Claim 7: Bienek, North, and Masako disclose the audio signal processing method according to claim 5, and wherein: an over-sampling period of the over-sampling operation is 1/N (N is an integer larger than or equal to 2) of the sampling period of the digital signal; and when the delay time represented by the decimal part is nearly an integral multiple (m) of the over-sampling period, m/N is adopted as the decimal part (Masako, Paragraph 26).

Claim 8: <u>Bienek</u>, <u>North</u>, and <u>Masako</u> disclose the audio signal processing method according to claim 7, wherein: the pulse-waveform data to be delayed by a delay time which is m/N (m = 1 to N - 1) of the sampling period is pre-stored in a data base; and pulse-waveform data approximate to the decimal part is taken out of the stored pulse-waveform data and set as a filter factor of each of the plurality of digital filters (<u>Bienek</u>, stored sets of delays (for the DDGs) and filter coefficients (for the ADFS) can be recalled, Page 14 Lines 26 and 27).

Claim 9: <u>Bienek</u>, <u>North</u>, and <u>Masako</u> disclose the audio signal processing method according to claim 5, wherein a transfer characteristic providing a predetermined acoustic effect is convoluted in the pulse-waveform data and set as a filter factor of each of the plurality of digital filters ("convolution multiplier", <u>Masako</u>, Paragraph 26).

Claim 14: Bienek discloses the audio signal processor according to claim 10, wherein: the pulse-waveform provided by the calculation circuit is set as a filter factor of each of the plurality of digital filters (Third Sound Field, Pages 21 – 22 and Figure 7C) but does not explicitly disclose the predetermined delay time set for at least one of the plurality of digital filters is divided into an integer part and decimal part in units of a sampling period of the audio signal, there is further provided a calculation circuit to calculate pulse-waveform data of the sampling period by over-sampling an impulse response including a delay time represented by at least the decimal part of the predetermined delay time

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for a shorter period than the sampling period to provide a sample train, and downsampling the sample train. Bienek does disclose "the minimum delay possible for a given signal is preferably as small or smaller than Ts, that signal's sample period" and that "most preferably, the smallest incremental change in delay possible for a given digital signal should be no larger than Ts, that signal's sampling period. Otherwise, interpolation of the signal is necessary," Page 12 Lines 17 – 24. Therefore Bienek does disclose a fractional delay and also discloses that a delay filter and an adaptive digital filter may be used. Bienek does not disclose details of how to perform, for example, the interpolation necessary for the fractional delays disclosed above and therefore one would be inclined to look elsewhere for such a teaching. Masako discloses the technique necessary to include a fractional delay (Figure 6 –7 and Paragraph 29 – 33) and discloses that this technique is an effective approach when slight spacing differences influence the felling of the direction of perceived sound. Therefore, it would have been obvious to one of ordinary skill in the art at the time of the invention to include the process of over-sampling an impulse response and then down-sampling to provide a pulse-waveform data of the sampling period as disclosed by Masako for setting the delay coefficients for the plurality of digital filters as disclosed by Bienek and North, thereby providing better perceived spatial resolution.

Claim 15: Bienek, North, and Masako disclose the audio signal processor according to claim 14, wherein: an over-sampling period of the over-sampling in the calculation circuit is 1/N (N is an integer larger than or equal to 2) of the sampling period of the

digital signal; and when the delay time represented by the decimal part is nearly an integral multiple (m) of the over-sampling period, m/N is adopted as the decimal part (Masako, Paragraph 26).

Claim 16: Bienek, North, and Masako disclose the audio signal processor according to claim 14, wherein a transfer characteristic providing a predetermined acoustic effect is convoluted in the pulse-waveform data to set synthetic-waveform data as a filter factor of each of the plurality of digital filters ("convolution multiplier", Masako, Paragraph 26).

Claim 17: Bienek discloses the audio signal processor according to claim 10, wherein: the pulse-waveform data stored in the storing means is taken out and set as a filter factor of each of the plurality of digital filters (stored sets of delays (for the DDGs) and filter coefficients (for the ADFS) can be recalled, Page 14 Lines 26 and 27) but does not explicitly disclose the predetermined delay time set for at least one of the plurality of digital filters is divided into an integer part and decimal part in units of a sampling period of the audio Signal; there is further provided a storing means for storing pulse-waveform data of the sampling period provided by over- sampling an impulse response including a delay time represented by at least the decimal part of the predetermined delay time for a shorter period than the sampling period to provide a sample train, and down-sampling the sample train. Bienek does disclose "the minimum delay possible for a given signal is preferably as small or smaller than Ts, that signal's sample period" and that "most preferably, the smallest incremental change in delay possible for a given digital signal

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should be no larger than Ts, that signal's sampling period. Otherwise, interpolation of the signal is necessary," Page 12 Lines 17 – 24. Therefore Bienek does disclose a fractional delay and also discloses that a delay filter and an adaptive digital filter may be used. Bienek does not disclose details of how to perform, for example, the interpolation necessary for the fractional delays disclosed above and therefore one would be inclined to look elsewhere for such a teaching. Masako discloses the technique necessary to include a fractional delay (Figure 6 –7 and Paragraph 29 – 33) and discloses that this technique is an effective approach when slight spacing differences influence the felling of the direction of perceived sound. Therefore, it would have been obvious to one of ordinary skill in the art at the time of the invention to include the process of oversampling an impulse response and then down-sampling to provide a pulse-waveform data of the sampling period as disclosed by Masako for setting the delay coefficients for the plurality of digital filters as disclosed by Bienek and North, thereby providing better perceived spatial resolution.

Claim 18: Bienek, North, and Masako disclose the audio signal processor according to claim 17, wherein: an over-sampling period of the over-sampling is 1/N (N is an integer larger than or equal to 2) of the sampling period of the digital signal; and when the delay time represented by the decimal part is nearly an integral multiple (m) of the over-sampling period, m/N is adopted as the decimal part (Masako, Paragraph 26).

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Claim 19: Bienek, North, and Masako disclose the audio signal processor according to claim 17, wherein: a plurality of the pulse-waveform data corresponding to the decimal part is pre-stored in the storing means; and pulse-waveform data approximate to the decimal part is taken out of the stored pulse-waveform data and set as a filter factor of each of the plurality of digital filters (Bienek, stored sets of delays (for the DDGs) and filter coefficients (for the ADFS) can be recalled, Page 14 Lines 26 and 27).

Claim 20: Bienek, North, and Masako disclose the audio signal processor according to claim 17, wherein a transfer characteristic providing a predetermined acoustic effect is convoluted in the pulse-waveform data to set the pulse-waveform data as a filter factor of each of the plurality of digital filters ("convolution multiplier", Masako, Paragraph 26).

Response to Arguments

7. Applicant's arguments with respect to claims 1 - 20 have been considered but are most in view of the new ground(s) of rejection in view of North (US 6,801,631).

Conclusion

8. Applicant's amendment necessitated the new ground(s) of rejection presented in this Office action. Accordingly, **THIS ACTION IS MADE FINAL**. See MPEP § 706.07(a). Applicant is reminded of the extension of time policy as set forth in 37 CFR 1.136(a).

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A shortened statutory period for reply to this final action is set to expire THREE MONTHS from the mailing date of this action. In the event a first reply is filed within TWO MONTHS of the mailing date of this final action and the advisory action is not mailed until after the end of the THREE-MONTH shortened statutory period, then the shortened statutory period will expire on the date the advisory action is mailed, and any extension fee pursuant to 37 CFR 1.136(a) will be calculated from the mailing date of the advisory action. In no event, however, will the statutory period for reply expire later than SIX MONTHS from the date of this final action.

Any inquiry concerning this communication or earlier communications from the examiner should be directed to Joseph Saunders whose telephone number is (571) 270-1063. The examiner can normally be reached on Monday - Thursday, 9:00 a.m. - 4:00 p.m., EST.

If attempts to reach the examiner by telephone are unsuccessful, the examiner's supervisor, Curtis Kuntz can be reached on (571) 272-7499. The fax phone number for the organization where this application or proceeding is assigned is 571-273-8300.

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Information regarding the status of an application may be obtained from the Patent Application Information Retrieval (PAIR) system. Status information for published applications may be obtained from either Private PAIR or Public PAIR. Status information for unpublished applications is available through Private PAIR only. For more information about the PAIR system, see http://pair-direct.uspto.gov. Should you have questions on access to the Private PAIR system, contact the Electronic Business Center (EBC) at 866-217-9197 (toll-free). If you would like assistance from a USPTO Customer Service Representative or access to the automated information system, call 800-786-9199 (IN USA OR CANADA) or 571-272-1000.

/J. S./
Examiner, Art Unit 2614
/CURTIS KUNTZ/
Supervisory Patent Examiner, Art Unit 2614